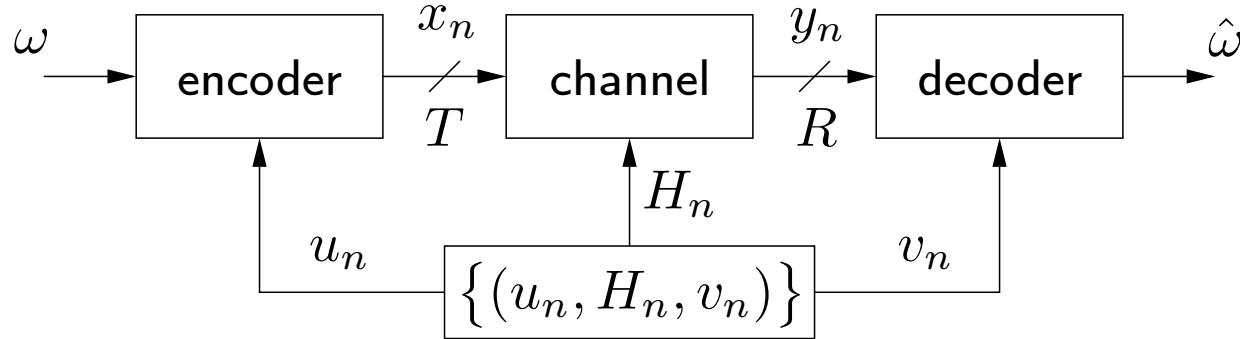


CAPACITY FORMULAS FOR MIMO LINKS

Mikael Skoglund
Signaler, sensorer och system
KTH

A WIRELESS MIMO LINK



- Equally likely information symbols $\omega \in \{1, \dots, M\}$
- T transmit and R receive antennas
- Channel matrix H_n
- CSIT u_n and CSIR v_n
- Received vector,
$$y_n = H_n x_n + w_n$$

where w_n is i.i.d Gaussian with $E[w_n w_n^*] = I_R$

- Coding over N uses of the T transmit antennas, subject to an average power constraint
 - ω mapped into (x_1, \dots, x_N) subject to
$$E\|x_n\|^2 \leq P$$
- Rate

$$R = \frac{\log M}{N} \quad [\text{bits per channel use}]$$
- Average probability of error: $P_e^{(N)} = \Pr(\hat{\omega} \neq \omega)$
- *Achievable rate*: R is achievable if $P_e^{(N)} \rightarrow 0$ possible as $N \rightarrow \infty$ at rate R
- *Capacity*: The capacity C is the supremum of all achievable rates

SCENARIOS

- **Time-variation of H_n**

- H.1 $H_n = H$, $n = 1, \dots, N$, where H is deterministic

- H.2 $H_n = H$, $n = 1, \dots, N$, where H is random

- H.3 $\{H_n\}$ stationary and ergodic (e.g., i.i.d)

- **CSIT**

- T.1 Perfect CSIT ($u_n = H_n$)

- T.2 No CSIT

- T.3 Partial CSIT ($u_n \neq H_n$ but depends on H_n)

- **CSIR**

- R.1 Perfect CSIR ($v_n = H_n$)

- R.2 No CSIT

- R.3 Partial CSIR ($v_n \neq H_n$ but depends on H_n)

CSIT and CSIR refer to *realizations* of H_n ; *statistics* of $\{H_n\}$ known

STATIC CHANNEL, PERFECT CSI

- **H.1, T.1 and R.1**

- $y_n = Hx_n + w_n$
- SVD of H : $H = Q_1 \Lambda Q_2^*$ \implies channel model equivalent to

$$\tilde{y}_n = Q_1^* y_n = \Lambda (Q_2^* x_n) + (Q_1^* w_n) = \Lambda \tilde{x}_n + \tilde{w}_n$$

- Non-zero singular values of H : $\lambda_1 \geq \lambda_2 \geq \dots \geq \lambda_r > 0$,
 $r \leq \min\{T, R\} \Rightarrow r$ parallel Gaussian channels

$$\tilde{y}_i = \lambda_i \tilde{x}_i + \tilde{w}_i, \quad i = 1, \dots, r$$

- Power allocation by *waterfilling*: Allocate power

$$P_i = E[\tilde{x}_i^2] = [\mu - \lambda_i^{-2}]^+$$

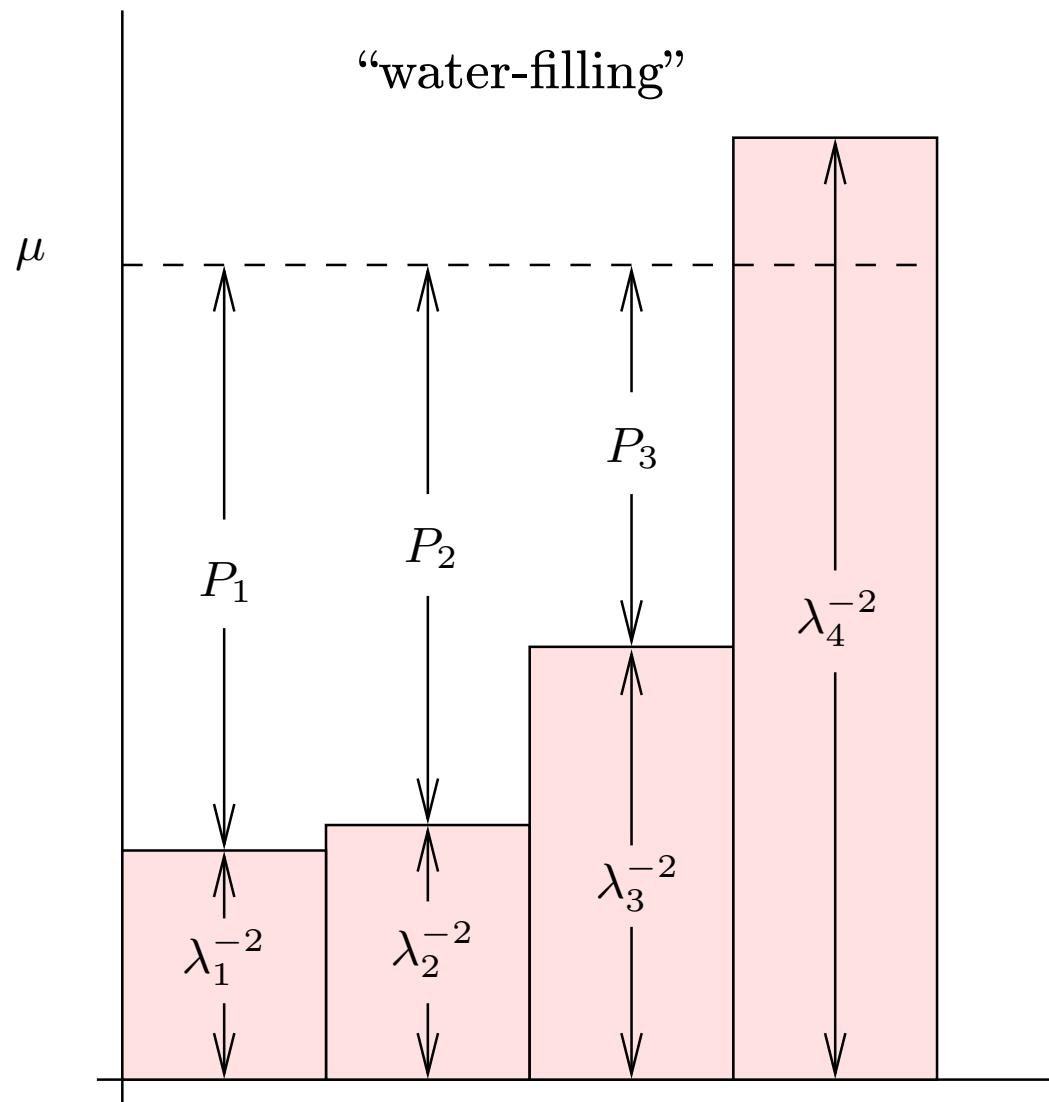
to channel i , with μ chosen so that

$$\sum_{i=1}^r P_i = P$$

- The capacity is then

$$C = \sum_{i=1}^r [\log(\mu\lambda_i^2)]^+$$

- H constant in time \Rightarrow *static* power allocation



- **H.2, T.1 and R.1**

- Non-zero singular values of H : $\lambda_1 \geq \lambda_2 \geq \dots \geq \lambda_r > 0$,
 $r \leq \min\{T, R\} \Rightarrow r$ parallel Gaussian channels

$$\tilde{y}_i = \lambda_i \tilde{x}_i + \tilde{w}_i, \quad i = 1, \dots, r$$

- Waterfilling, μ chosen so that $\sum_{i=1}^r [\mu - \lambda_i^{-2}]^+ = P$
- Let

$$f = \sum_{i=1}^r [\log(\mu \lambda_i^2)]^+$$

then

$$C = \sup\{\alpha : \Pr(f \leq \alpha) = 0\}$$

- Example: Components of H i.i.d complex Gaussian $\Rightarrow C = 0$

- **References**

- Static deterministic channel: classic result (e.g., Cover & Thomas)
- Static random channel, e.g.
 - * S. Verdú and T. S. Han, “A general formula for channel capacity,” *IEEE T-IT*, July 1994
 - * E. Biglieri, G. Caire and G. Taricco, “Limiting performance of block-fading channels with multiple antennas,” *IEEE T-IT*, May 2001

TIME-VARYING CHANNEL, PERFECT CSI

- **H.3, T.1 and R.1**

- Singular values of H_n : $\lambda_{1n} \geq \dots \geq \lambda_{nq} \geq 0$, $q = \min\{T, R\}$
 $\Rightarrow q$ parallel Gaussian channels at time n

$$\tilde{y}_{in} = \lambda_{in} \tilde{x}_{in} + \tilde{w}_{in}, \quad i = 1, \dots, q$$

- Power allocation by waterfilling \Rightarrow

$$C = \sum_{i=1}^q E[\log(\mu \lambda_{in}^2)]^+$$

with μ chosen so that $\sum_{i=1}^q E[\mu - \lambda_{in}^{-2}]^+ = P$

- H_n varies in time \Rightarrow *dynamic* power allocation

- **References**

- A. J. Goldsmith and P. P. Varaiya, “Capacity of fading channels with channel side information,” *IEEE T-IT*, Nov. 1997
- M. Skoglund and G. Jöngren, “On the capacity of a multiple-antenna communication link with channel side information,” *IEEE J-SAC*, Apr. 2003

STATIC CHANNEL, PERFECT CSIR, NO CSIT

- **H.1, T.2 and R.1**

$$C = ??????$$

problem ill-defined

- Nonsense formulas

$$C = \log \det(HQH^* + I_R)$$

$$C = \max_Q \log \det(HQH^* + I_R)$$

$$C = \log \det(HH^* + I_R)$$

- **H.2, T.2 and R.1**

$$C = ??????$$

problem ill-defined

- Nonsense formulas

$$C = \log \det(HQH^* + I_R)$$

$$C = \max_Q \log \det(HQH^* + I_R)$$

$$C = \log \det(HH^* + I_R)$$

with or without expectations taken

- A “random” C , “average” C , . . . still more nonsense

- Make the problem well-defined by adopting *additional signaling constraints*, e.g., $E[x_n x_n^*] = Q$ (with $\text{Tr } Q = P$) \implies
 - **H.1, T.2 and R.1**

$$C = \log \det(HQH^* + I_R)$$

- **H.2, T.2 and R.1**

$$C = \sup \left\{ \alpha : \Pr(\log \det(HQH^* + I_R) \leq \alpha) = 0 \right\}$$

* H i.i.d Gaussian $\Rightarrow C = 0$

TIME-VARYING CHANNEL, PERFECT CSIR, NO CSIT

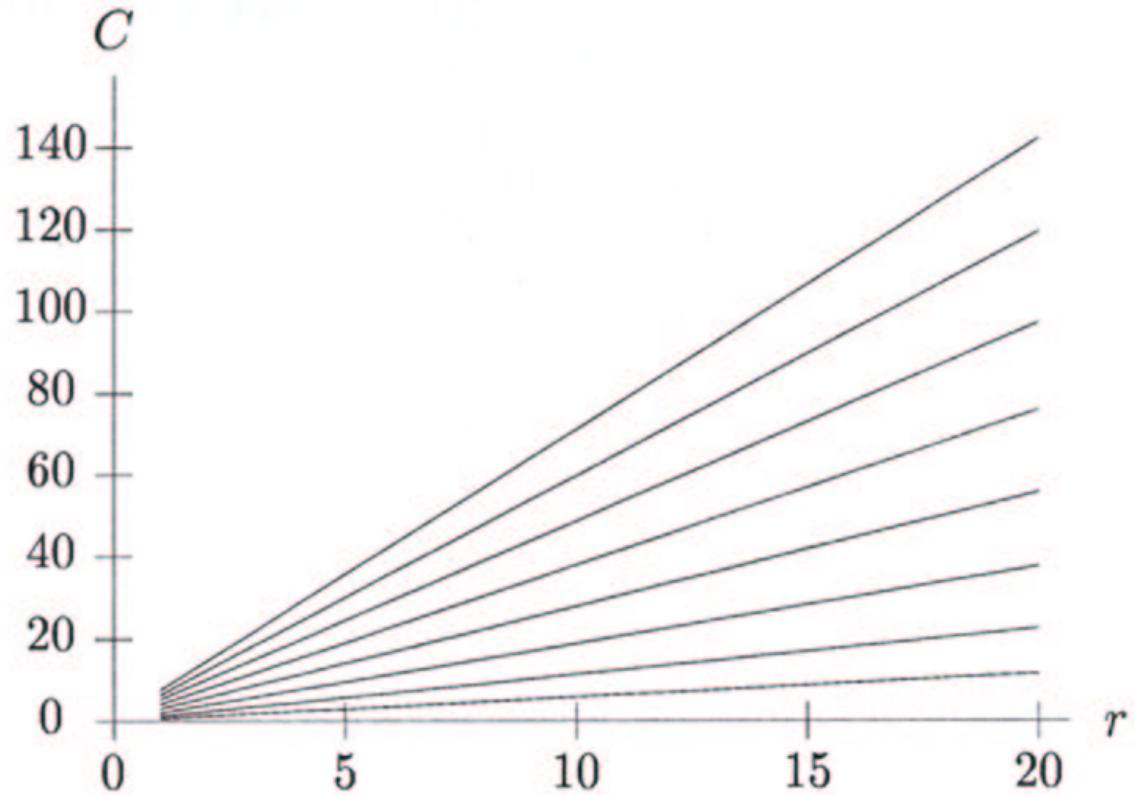
- **H.3, T.2 and R.1**

$$C = \max_{Q: \text{Tr } Q = P} E \log \det(HQH^* + I_R)$$

($H = H_n$ and over non-negative definite deterministic Q 's)

– $\{H_n\}$ i.i.d and elements of H_n i.i.d zero-mean Gaussian \Rightarrow

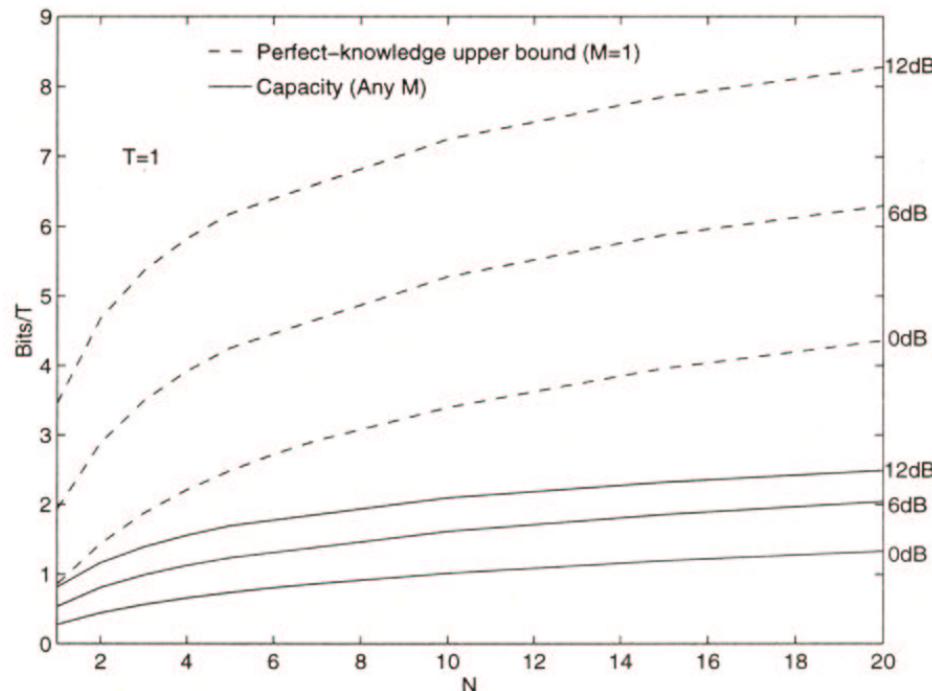
$$C = E \log \det(PHH^* + I_R), \quad \text{"Telatar's formula"}$$



C in nats from Telatar's formula; $T = R = r$; H_n i.i.d unit variance;
 $0 \leq P \leq 35$ [dB] in 5 dB increments

TIME-VARYING CHANNEL, NO CSI

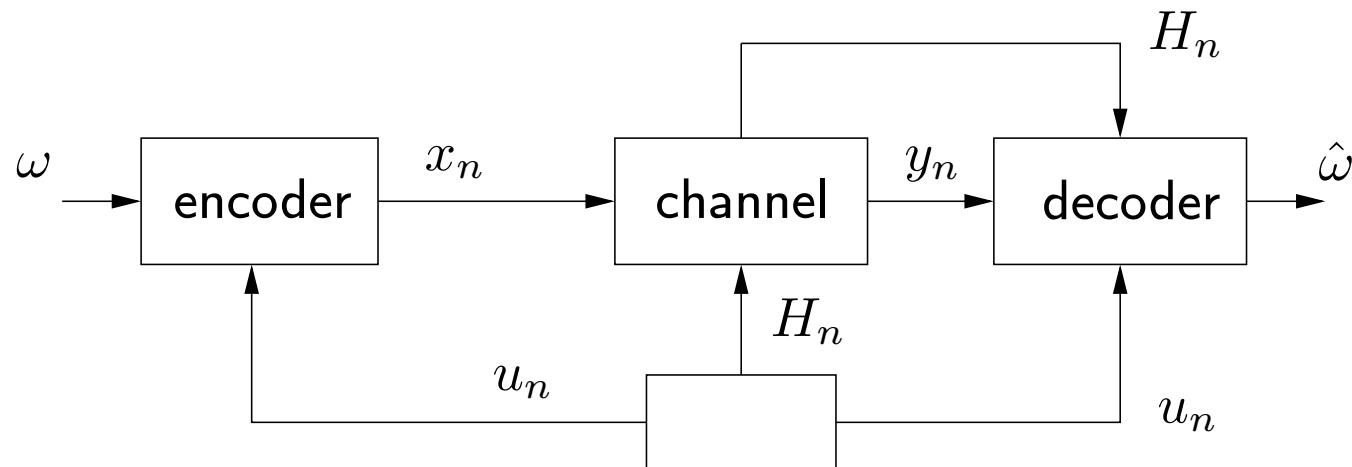
- **H.3, T.2 and R.2:** T. L. Marzetta and B. M. Hochwald, “Capacity of a mobile multiple-antenna communication link in Rayleigh flat fading,” *IEEE T-IT*, Jan. 99



H_n with i.i.d zero-mean Gaussian components. Any $T(M)$, $R = N$

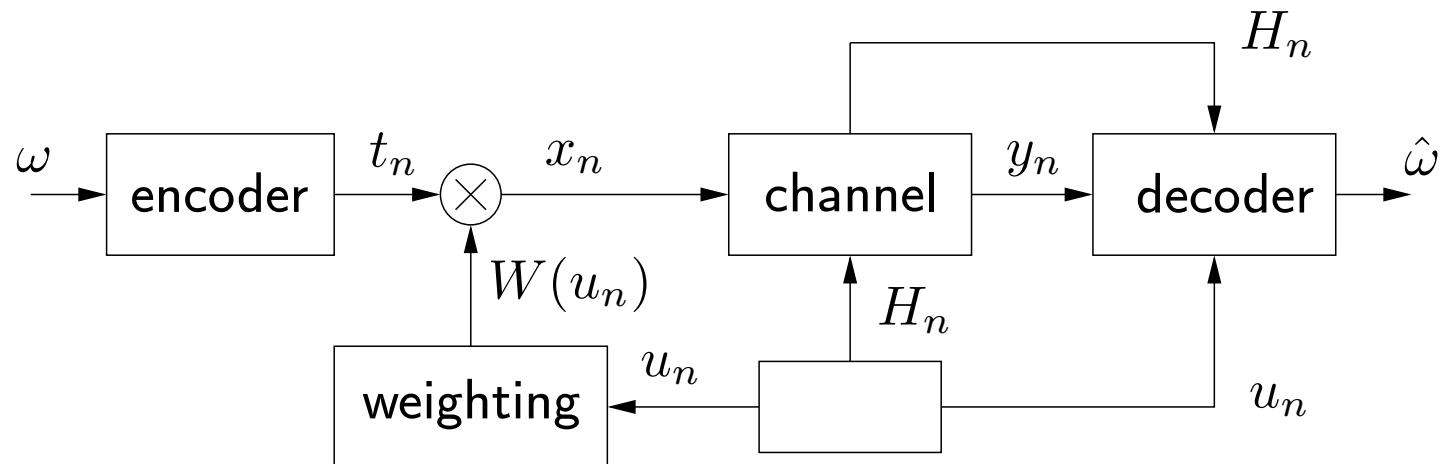
TIME-VARYING CHANNEL, PERFECT CSIR, PARTIAL CSIT

- **H.3, T.3 and R.1**, special case



- The receiver knows the value of u_n , e.g. deterministic feedback
- The above system has the same capacity as...

... this system

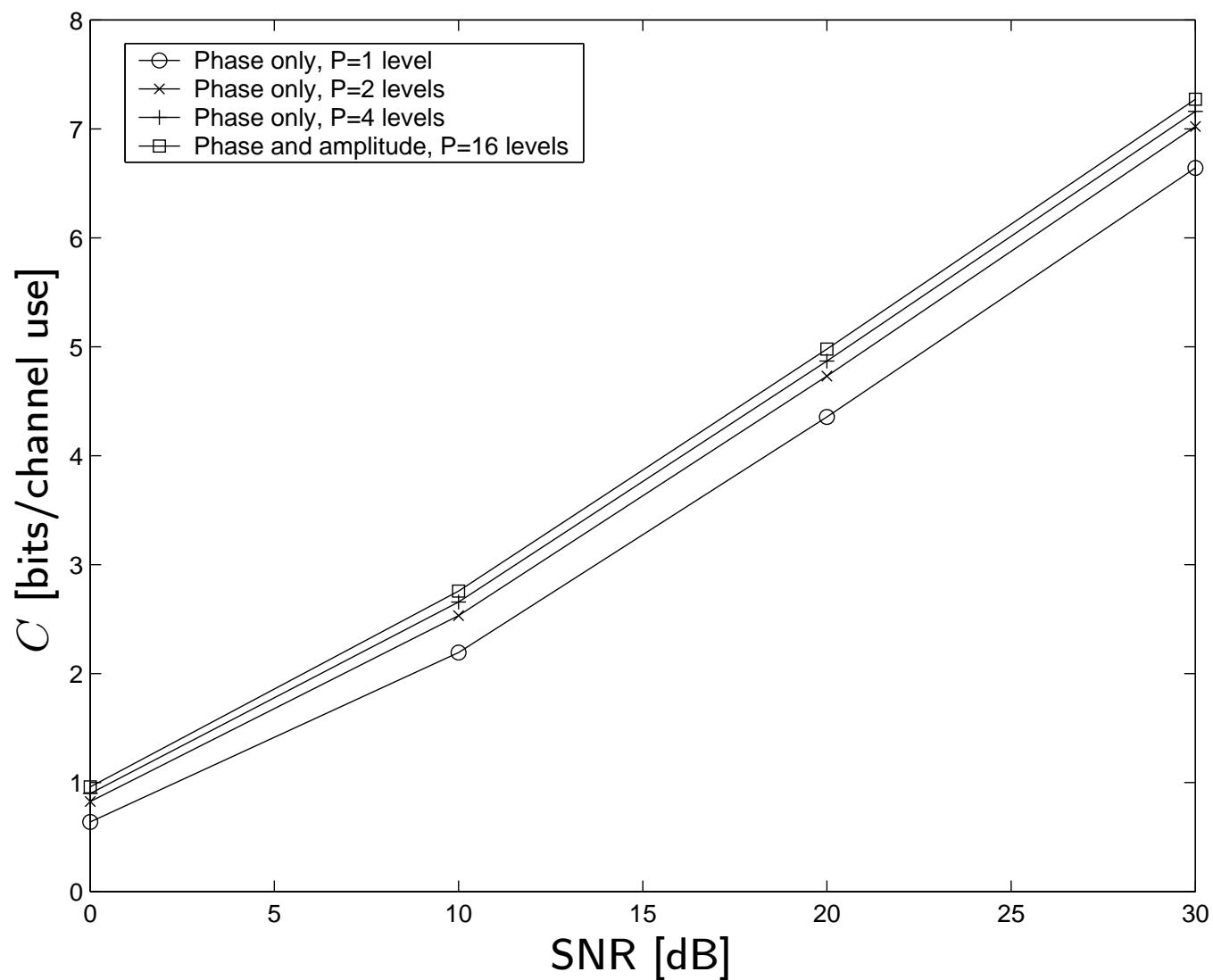


$$C = \max_{W : \text{Tr } E[WW^*] = P} E \log \det (H_n W(u_n) W^*(u_n) H_n^* + I_R)$$

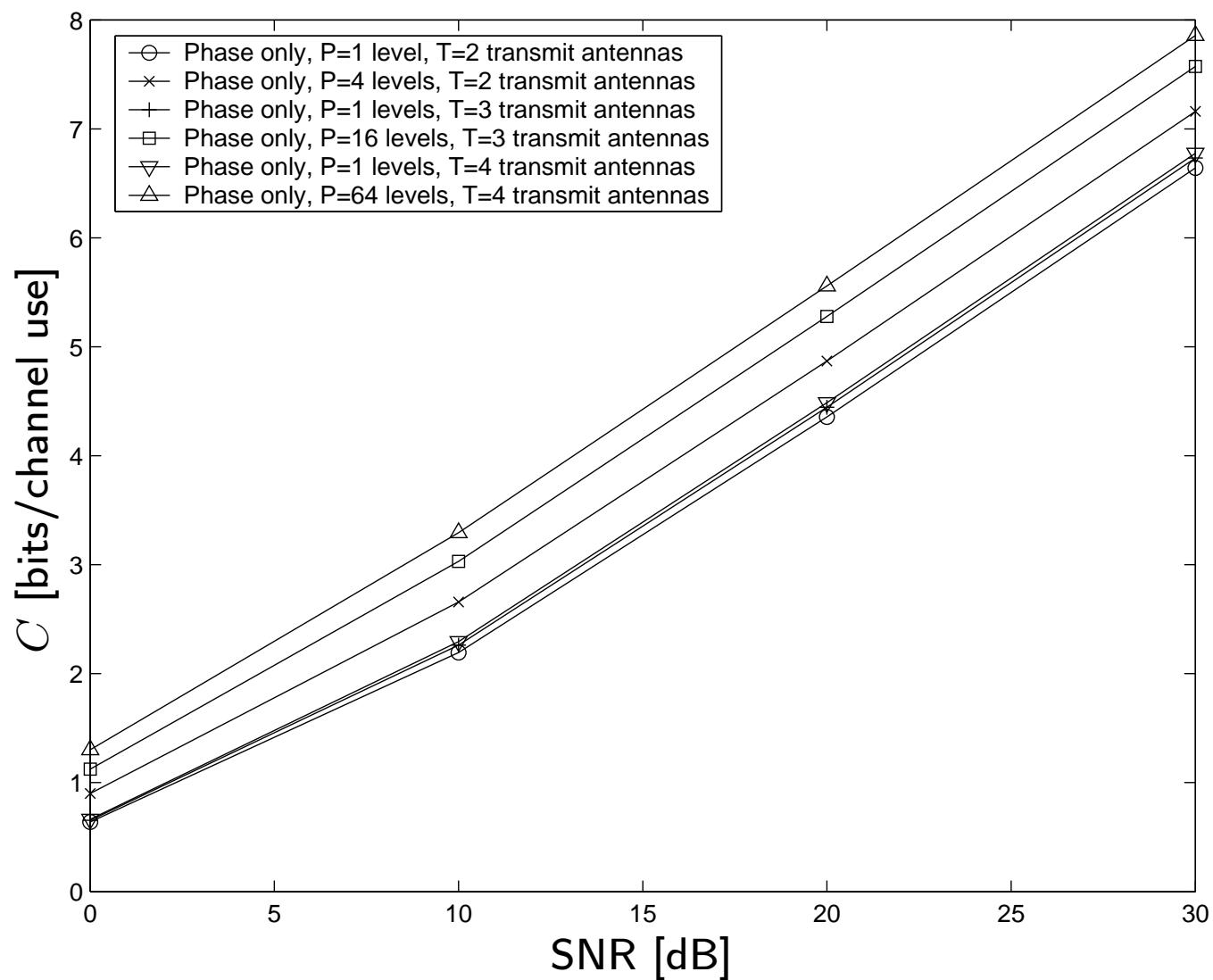
- Special cases,
 - Perfect CSIT (H.3, T.1, R.1): $u_n = H_n$
 - No CSIT (H.3, T.2, R.1): $u_n = \text{const}$

- $T = 2$ (transmit) and $R = 1$ (receive)
- H_n i.i.d Gaussian
- CSIR $v_n = H_n = [h_{1n} \ h_{2n}]$
- CSIT $u_n = P$ -valued quantized representation of H_n
 - *Phase only*: Uniform quantization of $\arg(h_{2n}/h_{1n})$
 - *Phase and amplitude* ($P = 16$): Quantize h_{2n}/h_{1n} to

$$s_i = \begin{cases} 0.8 e^{j i \pi / 4}, & i = 0, \dots, 7 \\ 1.6 e^{j (i-8) \pi / 4}, & i = 8, \dots, 15 \end{cases}$$



- T (transmit) and $R = 1$ (receive)
- H_n i.i.d. Gaussian
- CSIR $v_n = H_n = [h_{1n} \ \cdots \ h_{Tn}]$
- CSIT $u_n = P$ -valued quantized representation of H_n
 - Uniform quantization of $\arg(h_{ln}/h_{1n})$, $l = 2, \dots, T$



- **References**

- G. Caire and S. Shamai, “On the capacity of some channels with channel state information,” *IEEE T-IT*, Sept. 1999
- M. Skoglund and G. Jöngren, “On the capacity of a multiple-antenna communication link with channel side information,” *IEEE J-SAC*, Apr. 2003

SUMMARY & CONCLUSIONS

- Overview of capacity formulas, classified according to:
 - channel variation, CSIR and CSIT
- Important to know under what conditions your formula is valid,
 - some confusion exists in the literature
- Capacity is always a deterministic constant, never a random variable,
 - “average” capacity and “probability distributions” for capacity = nonsense
- Other information-theoretic measures of channel quality:
 - epsilon capacity, outage probability (supportable rate versus outage), coding exponents, . . .